

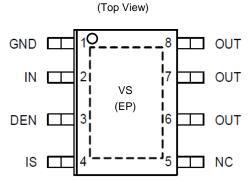
Description

The ZXMS81090SP is a single-channel, high-side power switch in a SO-8EP (Type E) exposed heatsink package incorporating protective and diagnostic functions.

The device comprises a monolithic n-channel vertical power MOSFET with integrated temperature and current sensors with a charge-pumped gate supply and has a low quiescent current in OFF state.

The device is enabled by active high 3.3V and 5V logic-level drive to the inputs. The device includes a diagnostic current-sense output proportional to load current and a defined diagnostic fault signal in case of overload operation, overtemperature, short-circuit or open- load conditions.

Pin Assignment



SO-8EP (Type E)

Features

Protection Functions

- Reverse Battery Protection Using External Components
- Voltage Dependent Current Limiting
- Overtemperature Protection with Auto-Restart
- Overvoltage Protection Including Load Dump
- Stable Undervoltage Protection
- ESD Protection
- Loss of Ground Protection with External Components
- Enhanced Short-Circuit Protection

Diagnostic Functions

- Proportional Load Current-Sense Output
 - Linear Voltage Drop Regulation to Maintain Sense Accuracy Even at Very Low Load Currents
 - Enabled by Logic Input
 - Defined Temperature and Current Dependency
- Open-Load Detection
 - Using Load Current-Sense in ON State
 - Using Output Voltage Detection in OFF State
- Defined Fault Signal in Case of Overload Operation,
 Overtemperature, or Short-Circuit and Open-Load in OFF State

Applications

- High-side switching with diagnostic feedback for:
 - 12V grounded loads
 - Resistive, inductive and capacitive loads
- Suitable for high inrush current loads
 - Incandescent lamps, motors, LEDs, etc.
- Compact low power replacement for:
 - Relays, fuses and discrete circuits

Summary Specifications

Parameter	Symbol	Rating
Operating Voltage	Vs	5V to 28V
Maximum Supply Voltage	V _{S(LD)}	41V
Maximum ON Resistance, T _J = +150°C	RDS(ON)	200mΩ
Nominal Load Current	I _L (NOM)	3A
Typical Current-Sense Ratio	Kılıs	1480
Minimum Current Limitation	I _{L5(SC)}	15A
Maximum Standby Current, T _J = +25°C	Is(OFF)	0.5μΑ

Miscellaneous

- Lead-Free Finish; RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- An automotive-compliant part is available under separate datasheet (ZXMS81090SPQ)

Notes:

- 1. EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant. All applicable RoHS exemptions applied.
- 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
- 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.

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Typical Applications Circuit

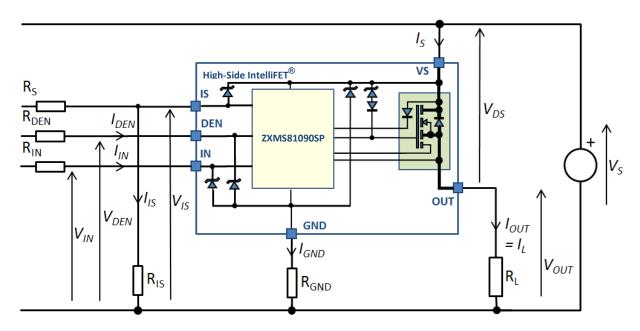


Figure 1. Typical Application Configuration

RGND, RIN, RDEN and Rs are optional. They may be replaced by shorts depending on the application. Non-zero resistors may be used to:

- Reduce peak currents during supply voltage transients that exceed the ±Vs internal clamp voltages
 - Typically transients exceeding 41V or may activate the internal clamps
- Protect the customer's application from high currents during transients exceeding 41V
- Keep within rated current during reverse battery, recommended is Rs = R_{DEN} = R_{IN} = 4.7kΩ, R_{GND} = 150Ω
- Ensure that the device is off when there is loss of ground connection to the device or module

Pin Description

Pin Number	Pin Name	Function and Description
1	GND	Ground or negative supply
2	IN	Input, activates the output
3	DEN	Diagnostic enable, allows common connection of the IS pin with multiple devices
4	IS	Diagnostic output, provides an analog sense current proportional to the load current under normal operation, or a defined current under overload or shutdown conditions
5	NC	Not connected
6, 7, 8	OUT	Output to the load, must be connected together
EP	VS	Voltage supply or battery positive

Table 1. Pin Description



Functional Block Diagram

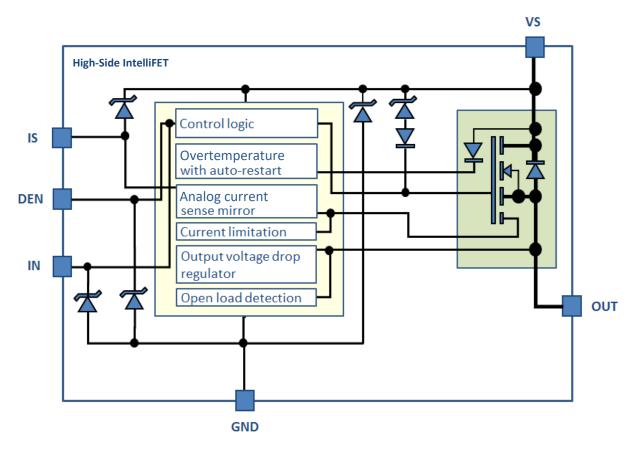


Figure 2. Functional Block Diagram of ZXMS81090SP



Absolute Maximum Ratings (Notes 4, 5) (@TJ = +25°C, unless otherwise specified.)

O mula a l	P	Rat	tings	11-14	O an allelana	
Symbol	Parameter	Min	Max	Unit	Conditions	
Supply Voltag	ge					
Vs	Supply voltage	-0.3	28	V	_	
V _{S(REV)}	Reverse supply voltage	0	16	٧	$R_L \ge 6\Omega$, $R_{GND} = 150\Omega$ $T_{amb} = +25$ °C, $t < 2$ mins	
Vs(sc)	Supply voltage for short-circuit protection (Note 6)	0	24	V	_	
V _{S(LD)}	Supply voltage for load-dump protection (ISO 7637)	_	41	V	$R_{IN} = 2\Omega$, $R_L = 6\Omega$	
Interface Pins	•					
\ /	IN nin voltage	-0.3	6	V	_	
Vin	IN pin voltage	_	7	V	t < 2 mins	
I _{IN}	Current in IN pin	-2	2	mA	_	
V	DEN sin veltage	-0.3	6	V	_	
VDEN	DEN pin voltage	_	7	V	t < 2 mins	
IDEN	Current in DEN pin	-2	2	mA	_	
V _{IS}	IS pin voltage	-0.3	Vs	V	_	
lıs	Current in IS pin	-25	50	mA	_	
Output Stage						
ال	Load current (Note 7)	_	Self-limited	Α	_	
Ртот	Power dissipation	_	1.6	W	T _{amb} = +85°C, T _J < +150°C	
Eas	Energy dissipation (single pulse)	_	40	mJ	Vs = 13.5V, IL = 3A T _J = +150°C	
VDS	VS to OUT pin voltage	_	41	V	_	
Nrsc	Repetitive short-circuit capability (Note 8)	_	300	kcyc	ton = 300ms	
Current						
1	Current in CND nin	-10	10	mA	_	
I _{GND}	Current in GND pin	-150	20	mA	t < 2 mins	
Temperature						
TJ	Junction temperature	-40	+150	°C	_	
Tstg	Storage temperature	-55	+150	°C	_	
Electrostatic	Discharge					
\/====	ESD capability HBM (all pins)	-2	2	kV	EIA/IESD 22 A 444D	
VESD(HBM)	ESD capability HBM OUT to GND and Vs shorted	-4	4	kV	- EIA/JESD 22-A 114B	
VESD(CDM)	ESD capability CDM	-0.75	0.75	kV	AEC-Q100-011	

Notes:

^{4.} Stresses greater than those listed under *Absolute Maximum Ratings* can cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to *Absolute Maximum Ratings* for extended periods can affect device reliability.

^{5.} Not subject to production test, guaranteed by design.

^{6.} Short-circuit protection is outside normal operation and is limited to single pulse and allows combinations of resistance and inductance.

^{7.} Current limit is a protection feature and operation in current limitation, e.g. with short-circuit loads, is outside the normal operation range.

^{8.} Repetitive short-circuit protection characterisation also carried out according to AEC-Q100-012 at 14V.



Package Thermal Data

Symbol	Parameter		Тур	Max	Unit
R _θ JC	Thermal resistance, junction-to-case (Note 9)		5.2		°C/W
R _θ ЈА	Thermal resistance, junction-to-ambient mounted on PCB (Notes 5, 9)		42	_	°C/W

Note:

Recommended Operating Conditions

Symbol	Parameter	Min	Max	Unit
V _{S(NOM)}	Nominal operating voltage (Note 10)	8	18	V
V _{S(OP)}	Extended operating voltage (Note 11)		28	V

Notes:

Operational Electrical Characteristics

(Unless otherwise specified: -40° C < T_J < $+150^{\circ}$ C; typical values based on T_J = $+25^{\circ}$ C)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
General						
Vs(OP)	Extended operating voltage	V _{IN} = 4.5V, V _{DS} < 0.5V	5	13.5	28	V
V _{S(OP_MIN)}	Undervoltage restart	$V_{IN} = 4.5V$, $R_L = 6\Omega$	3.8	4.2	5	V
Vs(uv)	Undervoltage shutdown	$V_{IN} = 4.5V, V_{DEN} = 0V$ $R_{L} = 6\Omega$ 3 3.3		3.3	4.1	V
V _{S(UV_HYS)}	Undervoltage hysteresis	_	_	0.85	_	V
IGND	Operating current	V _{IN} = V _{DEN} = 5.5V, V _S = 18V Device in R _{DS(ON)}	_	2	6	mA
	Ctondhy ourrant	V_{IN} and V_{DEN} floating $V_{OUT} = 0V$, $V_S = 18V$ $T_J \le +85^{\circ}C$	_	0.1	0.5	μΑ
I _{S(OFF)}	Standby current	V _{IN} and V _{DEN} floating V _{OUT} = 0V, V _S = 18V T _J = +150°C	_	3.5	20	μΑ
Is(off_den)	Standby current with diagnostic pin active	V _{IN} floating, V _{OUT} = 0V V _S = 18V, V _{DEN} = 5.5V	_	1	_	mA

^{9.} Device mounted on vertical PCB, 2" x 2" x 1.6mm, FR4 with 2oz copper for all connections.

^{10.} For normal function and protection features.

^{11.} Operation across an extended range is possible but is load dependant – device may have reduced protection against faulty (overload or short-circuit) loads



Output ON-State Characteristics

The ON-state resistance $R_{DS(ON)}$ depends on the supply voltage V_S and junction temperature T_J .

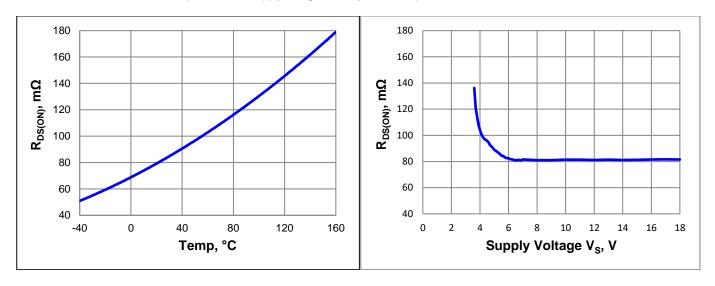


Figure 3. Typical R_{DS} ON-State Resistance

At low load current I_L , the MOSFET gate drive is reduced to maintain a near constant output voltage drop $V_{DS(NL)}$. This limits the effect of internal op-amp offset voltage, to maintain useful K_{ILIS} ratio accuracy even at very low I_L .

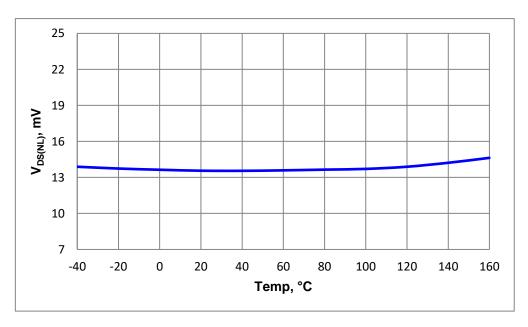


Figure 4. Output Voltage Drop Regulation



Resistive Load Switching

The power MOSFET turn-ON and turn-OFF processes are determined by the device itself, with rates suitable for EMC compatibility.

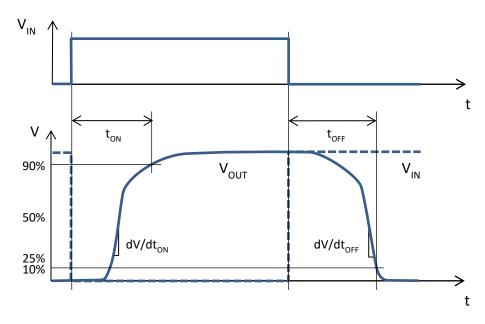


Figure 5. Switching a Resistive Load

Output Inductive Load Clamp

To de-energise inductive loads the OUT terminal must be allowed to swing below ground (VouT rings negative) during the OFF state.

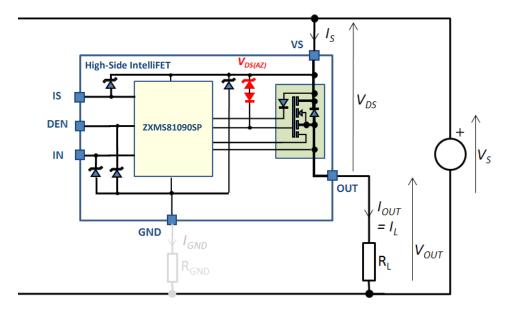


Figure 6. Output Clamp



Output Inductive Load Clamp (continued)

A low-impedance active voltage clamp uses the MOSFET channel to limit the maximum voltage across the MOSFET drain-source terminals, limiting the swing of OUT below V_S to safe $V_{DS(AZ)}$. This prevents avalanche of the MOSFET or associated circuitry.

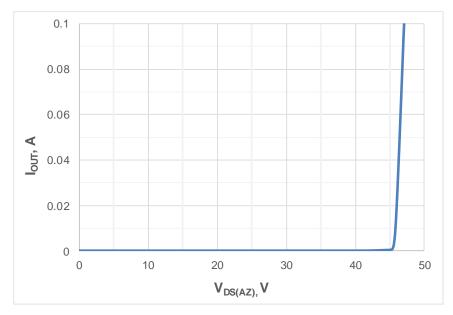


Figure 7. Typical V_{DS} Clamp Characteristic

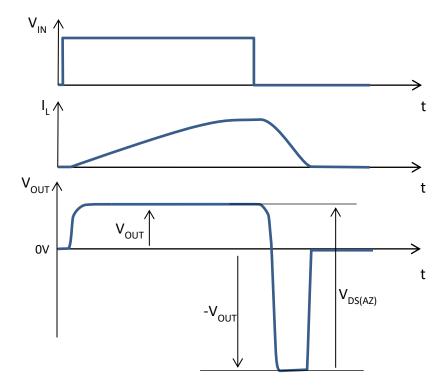


Figure 8. Switching an Inductive Load



Maximum Load Inductance

Stored inductive load ring-OFF energy is dissipated in the MOSFET during switching and load ring-OFF clamping. Additional energy is also supplied to the system by the Vs supply until the load current I_L reaches zero. This causes a temporary rise in MOSFET temperature after turn-OFF begins. The temperature reached depends on the starting temperature, thermal environment, load current I_L, load inductance L_L, load resistance R_L and supply voltage Vs.

Inverse Current Capability

In the ON-state the device will remain on if the output current becomes inverse until or unless the inverse current becomes high enough to create a -V_{DS} approaching body diode conduction. During inverse current conduction, I_{L(INV)}, the IS sense output will be zero. If body diode conduction occurs, all functions are disabled or unspecified until the inverse current becomes very small.

If inverse current is present in the OFF state body diode conduction occurs and all functions are disabled or unspecified. When inverse current is removed or becomes very small then turn-ON and normal function become possible.

Power Stage Electrical Characteristics

(Unless otherwise specified: $8V < V_S < 18V$, $-40^{\circ}C < T_J < +150^{\circ}C$; typical values based on $V_S = 13.5V$, $T_J = +25^{\circ}C$)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Output Charac	cteristics		•	•		
Rds(on)	ON-state resistance	V _{IN} = 4.5V, I _L = 4A T _J = +150°C		170	200	mΩ
()		V _{IN} = 4.5V, I _L = 4A		90		mΩ
I _L (NOM)	Nominal load current (Note 12)	Tamb = +85°C, T _J < +150°C	_	3	_	Α
VDS(NL)	Voltage drop regulation at low I∟	I _L = 50mA	_	12	25	mV
V _{DS(AZ)}	Output clamp voltage	I _L = 20mA	41	47	53	V
		V _{IN} floating, V _{OUT} = 0V	_	0.1	0.5	μA
I _L (OFF)	Output leakage current	VIN floating, VOUT = 0V Tamb = +150°C	_	3	10	μA
I _{L(INV)}	Inverse output current (Note 13)	Vs < Vout	_	2.5	_	Α
Timings						
dV/dton	Slew rate ON, 30% to 70% Vs		0.1	0.15	0.5	V/µs
dV/dtoff	Slew rate OFF, 70% to 30% Vs		0.1	0.25	0.7	V/µs
ton	Turn-ON time to 90% V _S	D 00 V 40 5V	30	150	230	μs
toff	Turn-OFF time to 10% Vs	$R_L = 6\Omega, V_S = 13.5V$	30	150	300	μs
ton(delay)	Turn-ON delay to 10% Vs		10	35	100	μs
toff(delay)	Turn-OFF delay to 90% V _S		30	100	200	μs
Eon	Switch ON energy	$R_L = 6\Omega, V_S = 18V$ $V_{OUT} = 90\% V_S$		1.1	_	mJ
Eoff	Switch OFF energy	$R_L = 6\Omega$, $V_S = 18V$ $V_{OUT} = 10\% V_S$		0.8	_	mJ

Notes:

- 12. Device mounted on vertical 50mm x 50mm x 1.5mm FR4 single-sided PCB with 6cm² 2oz copper in free air.
- 13. ON-state reverse conduction, functional test only.



Protection Features

Loss of Ground Protection

The device will turn off in the case that the ground pin connection is lost and the load remains connected. It is recommended to use the high ohmic input resistors in the interface pins to ensure that the device is turned off by limiting the current in the paths from the ground pin, through the input and diagnostic enable ESD diodes, to the external driving circuits.

Undervoltage Protection

The device will not turn on if the Vs supply voltage is below the minimum operating voltage $V_{S(OP_MIN)}$ where protection functions may not be operational. If the device is already on then the supply voltage has to drop to below the undervoltage threshold $V_{S(UV)}$ to turn the output off. Figure 9 shows the undervoltage mechanism.

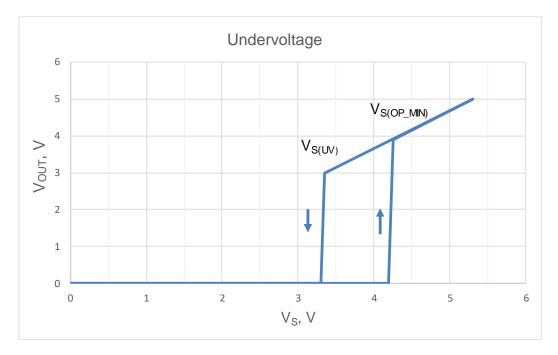


Figure 9. Undervoltage Behavior



Protection Features (continued)

Overvoltage Protection

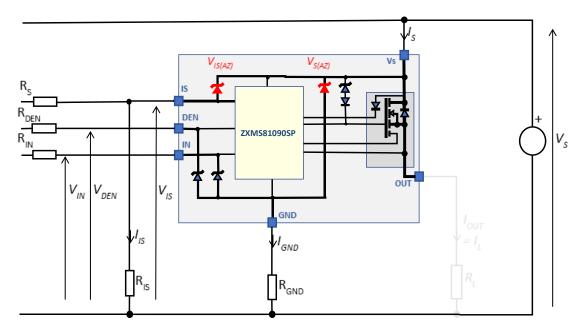


Figure 10. Overvoltage Clamping Circuit

The GND pin has an active protection clamp, operating much as a low noise high voltage Zener device, to protect it from overvoltage for high Vs transients. During Vs transient overvoltage the voltage is clamped and the excess voltage, Vs-Vs(AZ), is applied across the ground resistor R_{GND} raising the potential on the GND pin. Additional high ohmic series resistors may be needed to prevent high V_{IN} and V_{DEN} being applied directly to the driving circuits.

The IS pin also has an active protection clamp and during Vs transient overvoltage the voltage is clamped and excess voltage, Vs-V_{IS(AZ)}, is applied across the sense resistor R_{IS}. An additional high ohmic series resistor may be needed in the application to prevent high V_{IS} being applied directly to the application monitoring circuit.

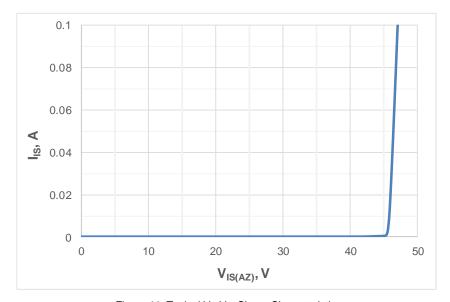


Figure 11. Typical V_S - V_{IS} Clamp Characteristic



Protection Features (continued)

Also during Vs overvoltage transient the voltage Vs-OUT is clamped and the excess voltage, Vs-V_{DS(AZ)}, is applied across the load R_L. Vs-OUT is the same clamp described under the earlier section <u>Output Inductive Load Clamp</u>.

Reverse Battery Protection

During reverse battery the output body diode is conducting current limited by the load itself resulting in power dissipation and the current in the ground and logic inputs has to be limited by external resistor components. No operating functions are available in this condition.

Overload Protection

During overload the output current is limited to a value depending on the V_{DS} voltage resulting in high dissipation in the output power stage. Sustained operation in this mode will raise the internal junction temperature until dynamic or absolute overtemperature protection cycling begins. There is a dynamic ($\Delta T_{J(SW)}$) and an absolute ($T_{J(SC)}$) temperature sensor. Figure 12 gives a sketch of overload protection.

If the temperature rise of the power stage versus the cooler control area exceeds $\Delta T_{J(SW)}$ then the device will be turned off until the rise falls to a reset level. Each cycle causes the absolute temperature to increase a little.

If the absolute temperature reaches $T_{J(SC)}$ then the device will be turned off until the absolute temperature falls by $\Delta T_{J(SC)}$. The device will continue to cycle to $T_{J(SC)}$ as long as the fault condition remains.

The IS pin outputs I_{IS(FAULT)} during load current limitation, during dynamic overtemperature cycling, and absolute overtemperature cycling.

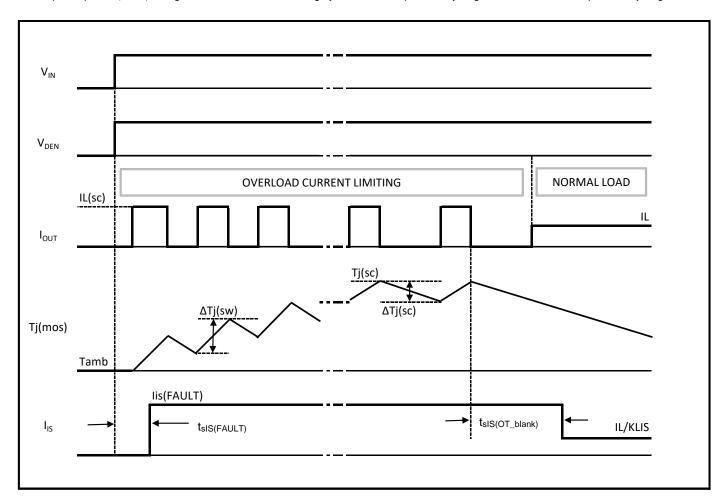


Figure 12. Overload Protection Diagram



Protection Electrical Characteristics

(Unless otherwise specified: $8V < V_S < 18V$, $-40^{\circ}C < T_J < +150^{\circ}C$; typical values based on $V_S = 13.5V$, $T_J = +25^{\circ}C$)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Loss of Gro	ound Protection					
I _{OUT} (GND)	Output leakage current (Note 14)	Vs = 28V GND disconnected	_	0.1	_	mA
Reverse Ba	ttery Protection					
V _{DS(REV)}	Reverse output voltage (Note 15)	I _L = -2A; T _J = +150°C	0.2	0.65	0.7	V
Overvoltag	e Protection					
V _{S(AZ)}	V _S to GND clamping	I _S = 5mA	41	47	53	V
Overload P	rotection					
I _{L5(SC)}	Load current limit	V _{DS} = 5V	15	23	30	Α
I _{L28(SC)}	Load current limit (Note 14)	V _{DS} = 28V	_	15	_	Α
I _L (RMS)	Load current during overtemperature cycling (Note 14)	$R_{SHORT} = 0.1\Omega$ LSHORT = 5µH	_	3	_	Α
ΔTJ(SW)	Dynamic temperature rise during cycling (Note 16)	_	_	80	_	K
T _J (SC)	Thermal shutdown temperature (Note 17)	_	+150	+170	+200	°C
$\Delta T_{J(SC)}$	Thermal hysteresis (Note 17)	_	_	20	_	K

Notes:

- 14. Not subject to production test, guaranteed by design.
- 15. During reverse battery the body-drain diode will conduct with voltage drop $V_{DS(REV)}$.
- 16. Functional test only.
- 17. Functional test only at T_J = +150°C.

Diagnostic Functionality - Detailed Description

In normal operation the IS pin outputs a small analog sense current proportional to the main OUT current flowing in the power MOSFET. In the case where it is disabled by the DEN pin it becomes high impedance. During overload/current limit operation/overtemperature/high temperature gradient or open load in the OFF state, the IS pin outputs a defined current I_{IS(FAULT)} greater than normal sense currents for normal loads. During ON-state operation with open load, normal OFF-state, or OFF-state with inductive load ring-off current still flowing, the IS current is approximately zero.

Operating Condition	Vin	V _{DEN}	Vout	IS Output Current, Is
Normal operation	L H	H H	Z H	Z = IL / KILIS
Current limiting	Н	Н	Н	lis(fault)
Short circuit OUT to GND	L H	H H	L L	Z IIS(FAULT)
Overtemperature	L H	H H	Z Z	Z I _{IS(FAULT)}
Short circuit OUT to Vs	L H	H H	H H	IIS(FAULT) < IL / KILIS (Note 18)
Open load	L L H	H H H	< VOL(OFF) > VOL(OFF) H	Z IIS(FAULT) < IIS(OL) (Note 19)
Inverse load current	L H	H H	H H	IIS(FAULT) IIS(OL)
All	X	L	Х	Z

Table 2. Operational Truth Table (Note 20)

Notes: 18. A low resistance short between OUT and V_S will reduce the output current, I_L and therefore reduce the analog sense current, I_{IS}.

- 19. With external pullup resistor.
- 20. H = high level; L = low level; Z = high impedance, voltage depends on external circuit; X = don't care.



Diagnostic Diagrams

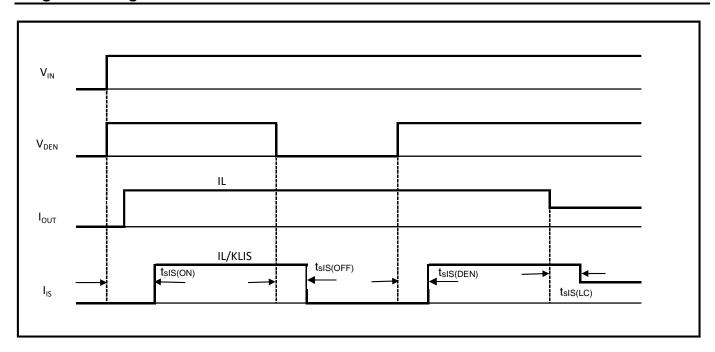


Figure 13. IS Signal Timing Diagram

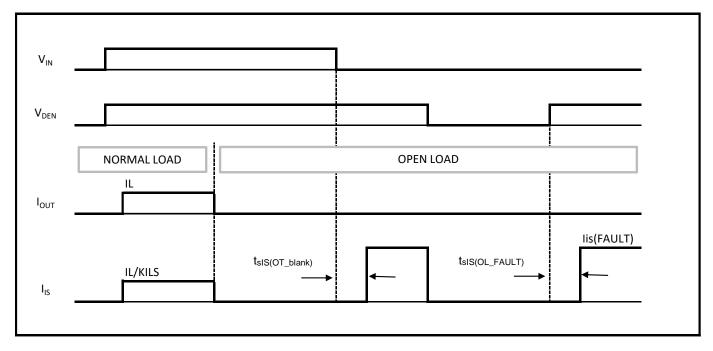


Figure 14. Open-Load Timing Diagram



Diagnostic Electrical Characteristics (Unless otherwise specified: $8V < V_S < 18V$, $-40^{\circ}C < T_J < +150^{\circ}C$; typical values based on $V_S = 13.5V$, $T_J = +25^{\circ}C$)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Open-Load	I Detection					
V _{DS(OL)}	Open-load OFF state detection voltage	V _{IN} = 0V, V _{DEN} = 4.5V	3.5	_	5.2	V
I _{L(OL)}	Open-load ON state detection current	$V_{IN} = V_{DEN} = 4.5V$ $I_{IS(OL)} = 8\mu A$	5	_	30	mA
Current Ser	nse Pin					
lis(DIS)	Current sense leakage current	$V_{IN} = 4.5V$, $V_{DEN} = 0V$ $I_L = 4A$	_	_	1	μА
V _{IS(SAT)}	VS to IS saturation voltage	$V_{IN} = 0V$, $V_{DEN} = 4.5V$ $I_{IS} = 6mA$, $V_{OUT} = V_S > 10V$	_	_	3	V
I _{IS(FAULT)}	Current sense under fault conditions	VIN = VIS = 0V, VDEN = 4.5V VOUT = VS > 10V	6	15	35	mA
VIS(AZ)	VS to IS clamp voltage	lis = 5mA	41	47	53	V
Load Curre	nt Sense					
K _{ILIS0}	Sense current ratio I _L = 50mA		-25%	1640	+25%	_
KILIS1	Sense current ratio I _L = 0.5A]	-16%	1480	+16%	_
K _{ILIS2}	Sense current ratio I _L = 1A		-10%	1480	+10%	_
K _{ILIS3}	Sense current ratio I _L = 2A	VIN = VDEN = 4.5V	-7%	1480	+7%	_
K _{ILIS4}	Sense current ratio I _L = 4A]	-6.5%	1480	+6.5%	_
ΔKILIS	Sense current ratio variation I _L = 2A versus I _L = 1A		-4%	_	+4%	_
Diagnostic	Timings					
tsIS(ON)	Current sense settling time to 90% I _{IS} after IN and DEN high	Vs = 13.5V, VIN = VDEN = 4.5V IL = 2A, RIS = 1.2kΩ	_	_	250	μs
tsIS(DEN)	Current sense settling time to 90% lis after DEN high	$V_S = 13.5V$, $V_{IN} = 4.5V$ $I_L = 2A$, $R_{IS} = 1.2k\Omega$	_	_	20	μs
tsIS(LC)	Current sense settling time to 90% lis after load current change	$V_S = 13.5V$, $V_{IN} = V_{DEN} = 4.5V$ $I_L = 1A$ to $2A$, $R_{IS} = 1.2k\Omega$	_	_	20	μs
t _{sIS(OL_}	Diagnostic fault current settling time to 90% IIS(FAULT) after DEN high with OFF state open- load condition	$V_{IN} = 0V$, $V_{OUT} = V_S = 13.5V$ R _{IS} = 1.2k Ω	_	_	150	μs
t _{sIS} (FAULT)	Diagnostic fault current settling time to 90% IIS(FAULT) after IN and DEN high with overload condition	V _{DS} = 5V, R _{IS} = 1.2kΩ			250	μs
tsIS(OT_ blank)	Diagnostic fault current off delay time to 90% I _{IS(FAULT)} after overtemperature condition returning to normal operation (Note 21)	V _{IN} = V _{DEN} = 4.5V R _{IS} = 1.2kΩ	_	150	_	μs
t _{sIS(OFF)}	Current sense fall time to < 50% lis after DEN low	$V_{IN} = 4.5V, I_{L} = 2A$ $R_{IS} = 1.2k\Omega$	_	_	30	μs

Note: 21. Not subject to production test, guaranteed by design.



Input Pins

The input circuit is compatible with 3.3V and 5V logic levels. The input diode provides ESD protection. If the pin is left open the internal tie down resistor will keep the output off. A Schmitt trigger provides switching hysteresis to avoid an undefined state if there is a slowly rising or falling voltage on the IN pin. Figure 15 shows the electrical equivalent circuit.

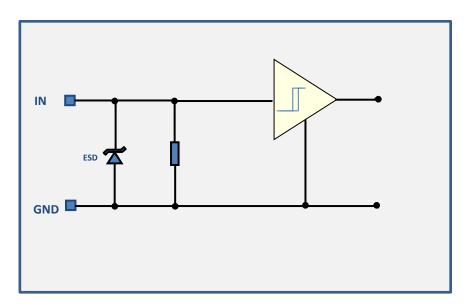


Figure 15. Input Pin Circuitry

The DEN pin has the same circuitry as the IN pin above.

Input Electrical Characteristics

(Unless otherwise specified: $8V < V_S < 18V$, $-40^{\circ}C < T_J < +150^{\circ}C$; typical values based on $V_S = 13.5V$, $T_J = +25^{\circ}C$)

Symbol	Parameter	Parameter Conditions		Тур	Max	Unit
IN Pin						
VIN(L)	Low level input voltage	_	-0.3	_	0.8	٧
V _{IN(H)}	High level input voltage	_	2	_	6	٧
V _{IN(HYS)}	Input voltage hysteresis	_	_	0.25	_	٧
lin(L)	Low level input current	V _{IN} = 0.8V	1	3	25	μΑ
lin(H)	High level input current	V _{IN} = 5.5V	2	8	25	μΑ
DEN Pin						
V _{DEN(L)}	Low level input voltage	_	-0.3	_	0.8	V
V _{DEN(H)}	High level input voltage	_	2	_	6	V
VDEN(HYS)	Input voltage hysteresis	_	_	0.25	_	V
I _{DEN(L)}	Low level input current	V _{DEN} = 0.8V	1	3	25	μA
IDEN(H)	High level input current	V _{DEN} = 5.5V	2	8	25	μA



Characterisation – General Product

Minimum Functional Supply Voltage

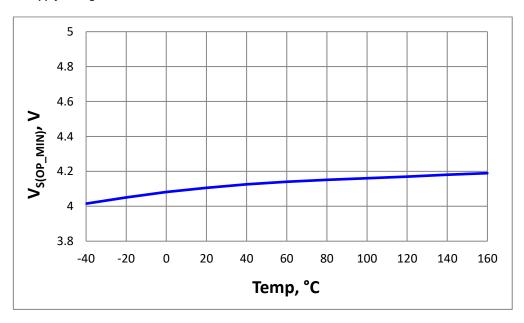


Figure 16. Minimum Functional Supply Voltage $V_{S(OP_MIN)} = f(T_J)$

Undervoltage Shutdown

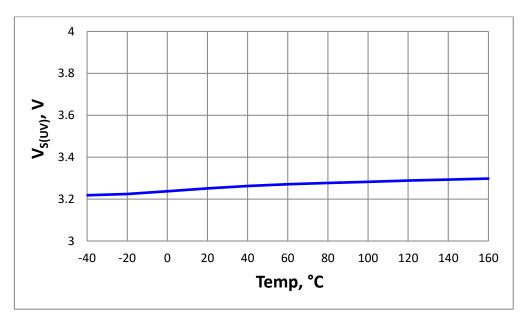


Figure 17. Undervoltage Shutdown $V_{S(UV)} = f(T_J)$



Characterisation - General Product (continued)

Current Consumption Channel Active

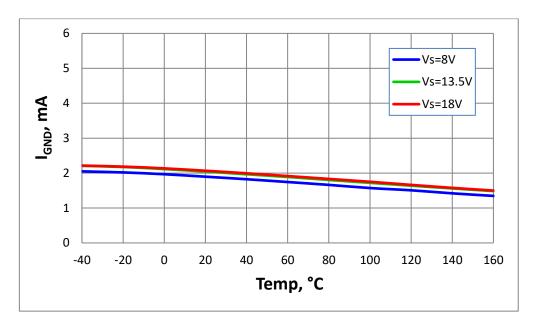


Figure 18. Current Consumption for Whole Device with Load, Channel Active I_{GND} = f(T_J;V_S)

Standby Current for Whole Device with Load

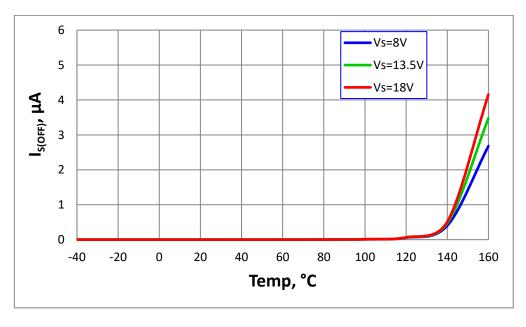


Figure 19. Standby Current for Whole Device with Load $I_{S(OFF)} = f(T_J; V_S)$



Characterisation - Power Stage

Output Voltage Drop Limitation at Low Load Current

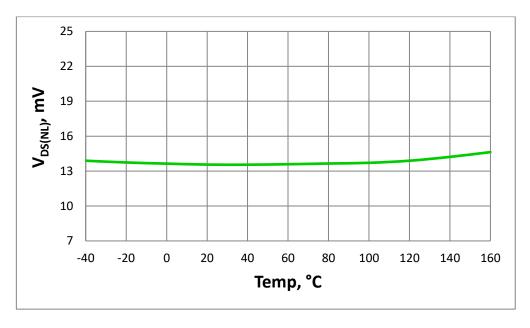


Figure 20. Output Voltage Drop Limitation at Low Load Current V_{DS(NL)} = f(T_J)

Drain to Source Clamp Voltage

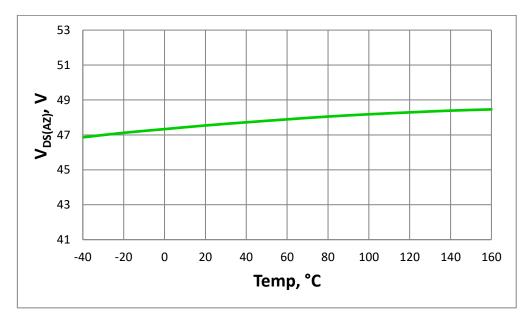


Figure 21. Drain to Source Clamp Voltage $V_{DS(AZ)} = f(T_J)$



Characterisation - Power Stage (continued)

Slew Rate at Turn ON

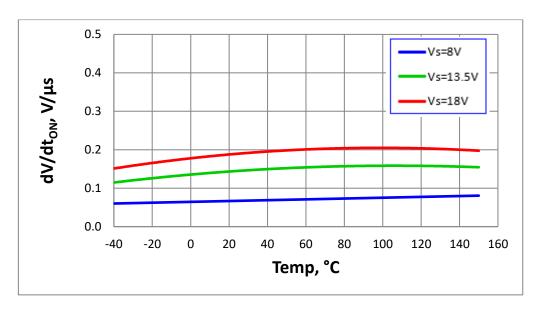


Figure 22. Slew Rate at Turn ON $dV/dtoN = f(T_J; V_S)$, $R_L = 6\Omega$

Slew Rate at Turn OFF

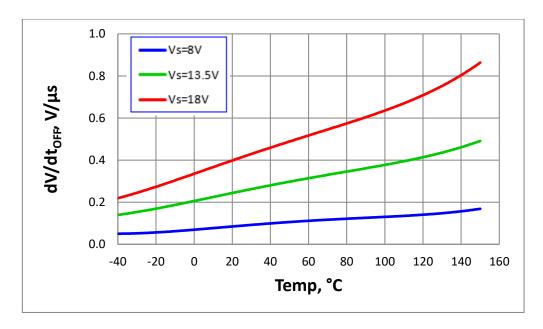


Figure 23. Slew Rate at Turn OFF dV/dtoFF = $f(T_J;V_S)$, R_L = 6Ω



Characterisation – Power Stage (continued)

Turn ON Time

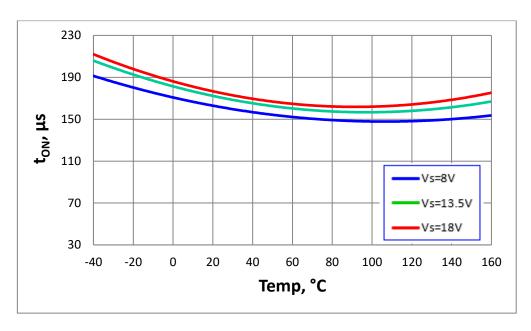


Figure 24. Turn ON ton = $f(T_J; V_S)$, $R_L = 6\Omega$

Turn OFF Time

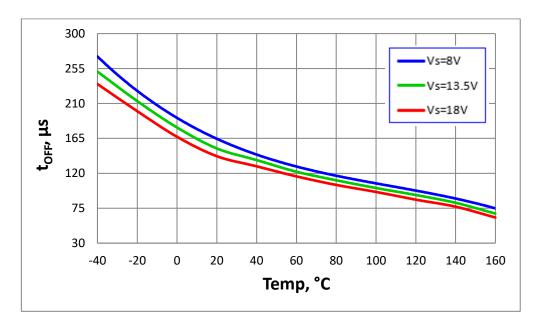


Figure 25. Turn OFF to $FF = f(T_J; V_S)$, $R_L = 6\Omega$



Characterisation - Power Stage (continued)

Switch ON Energy

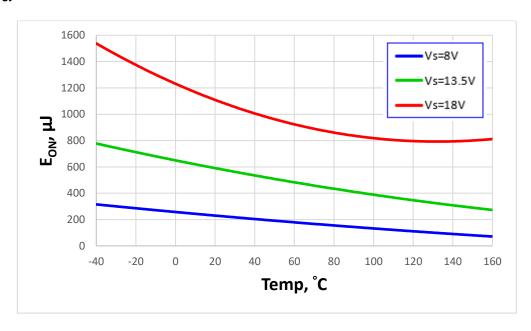


Figure 26. Switch ON Energy $E_{ON} = f(T_J; V_S)$, $R_L = 6\Omega$

Switch OFF Energy

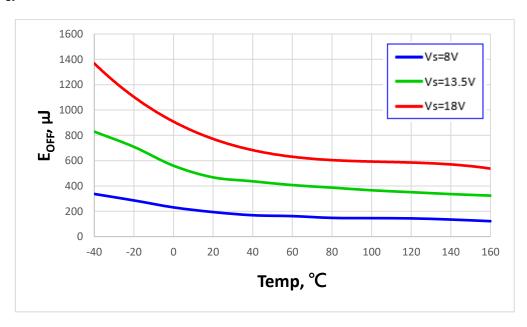


Figure 27. Switch OFF Energy $E_{OFF} = f(T_J; V_S), R_L = 6\Omega$



Characterisation - Protection

Overload Condition with Low Output Voltage Drop

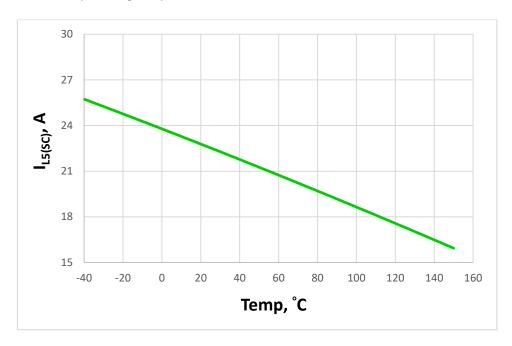


Figure 28. Overload Condition with Low Output Voltage Drop $I_{L5(SC)} = f(T_J)$

Overload Condition with High Output Voltage Drop

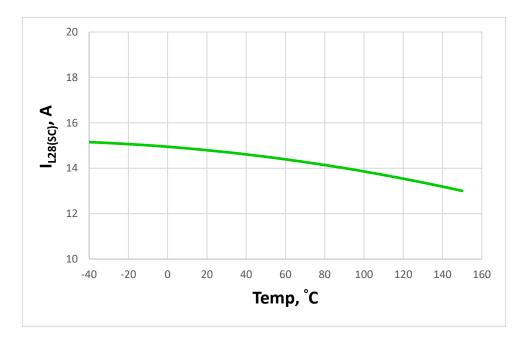


Figure 29. Overload Condition with High Output Voltage Drop $I_{L28(SC)} = f(T_J)$



Characterisation - Diagnostic Mechanism

Current Sense at No Load

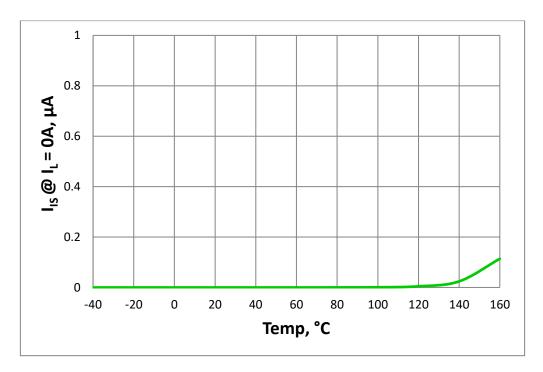


Figure 30. Current Sense at No Load I_{IS} = $f(T_J)$

Open-Load Detection Threshold in ON State

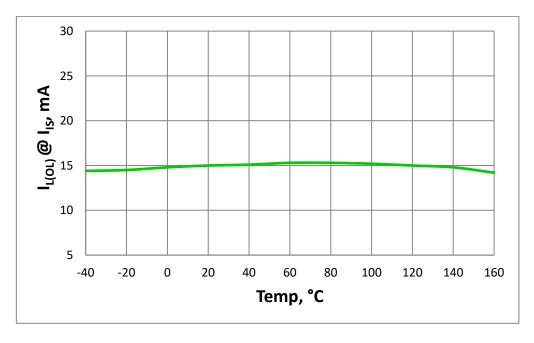


Figure 31. Open-Load Detection ON State Threshold $I_{L(OL)} = f(T_J)$



Characterisation - Diagnostic Mechanism (continued)

Sense Signal Maximum Voltage

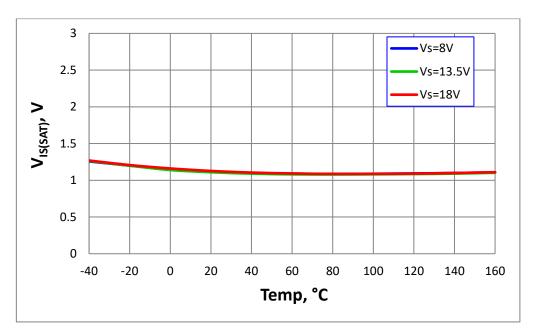


Figure 32. Sense Signal Maximum Voltage $V_{IS(SAT)} = f(T_{J};V_{S})$

Sense Signal Maximum Current

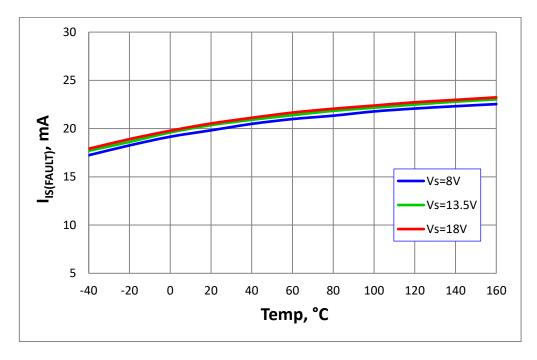


Figure 33. Sense Signal Maximum Current in Fault Condition $I_{IS(FAULT)} = f(T_J; V_S)$



Characterisation – Input Pins

Input Voltage Threshold ON to OFF

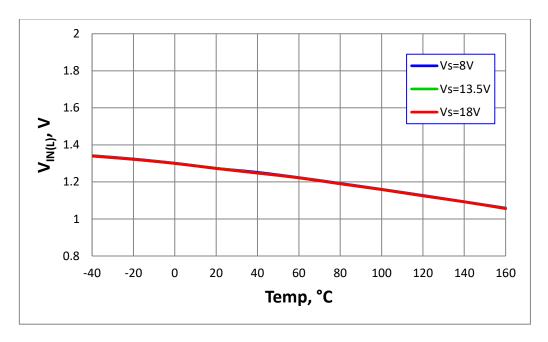


Figure 34. Input Voltage Threshold $V_{IN(L)} = f(T_J; V_S)$

Input Voltage Threshold OFF to ON

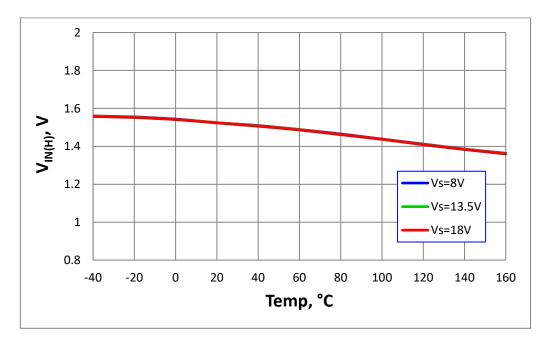


Figure 35. Input Voltage Threshold $V_{IN(H)} = f(T_J; V_S)$



Characterisation - Input Pins (continued)

Input Voltage Hysteresis

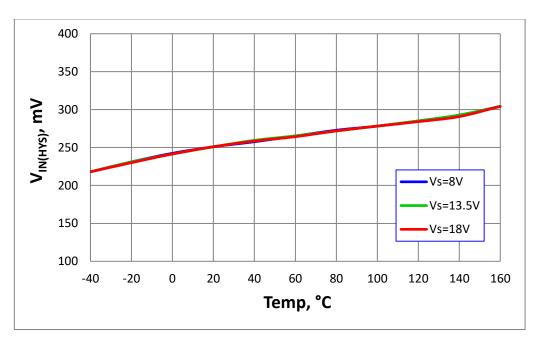


Figure 36. Input Voltage Hysteresis $V_{IN(HYS)} = f(T_J; V_S)$

Input Current High Level

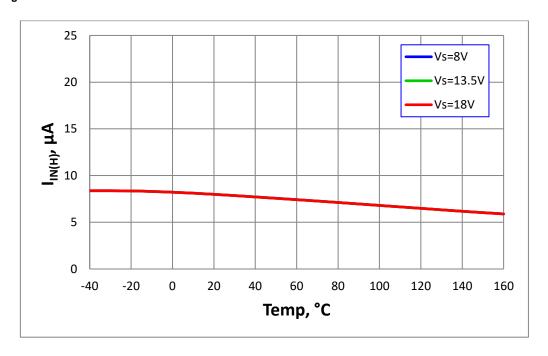


Figure 37. Input Current High Level $I_{IN(H)} = f(T_J; V_S)$



Application Information (Note 22)

Note: The following information is given as a hint for the implementation of the device only and shall not be regarded as a description or warranty of a certain functionality, condition or quality of the device.

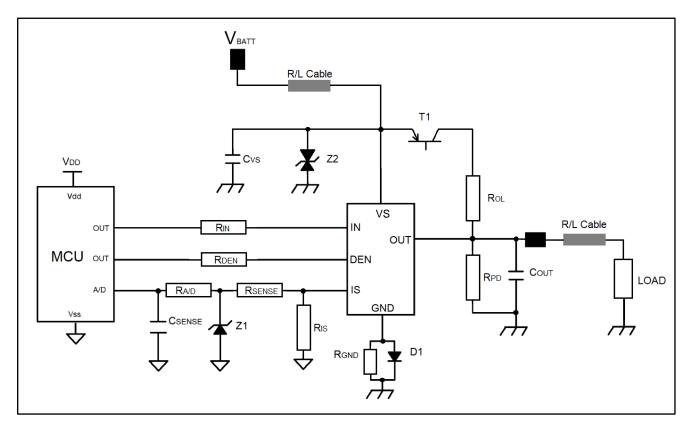


Figure 38. Application Circuit Diagram

Note: 22. This is a simplified example of an application circuit. The function must be verified in the real application.

Reference	Value	Purpose
R _{IN}	4.7kΩ	Connect to the micro controller for overvoltage, reverse polarity protections
R _{DEN}	4.7kΩ	Connect to the micro controller for overvoltage, reverse polarity protections
R _{PD}	47kΩ	Improve immunity to electromagnetic noise
R _{IS}	1.2kΩ	Sense resistor
RSENSE	4.7kΩ	Overvoltage, reverse polarity, loss of ground. Value to be tuned with micro controller specification
R _{OL}	1.5kΩ	For open load in OFF diagnostic
R _{A/D}	4.7kΩ	Protection for the micro controller during overvoltage, reverse polarity
R _{GND}	1kΩ	To keep the device GND at a stable potential during clamping
D1	BAS21	Protection of the device during reverse polarity
Z1	7V Zener diode	Protection of the micro controller during overvoltage
Z2	36V Zener diode	Protection of the device during overvoltage
T1	BC 807	Switch the battery voltage for open load in OFF diagnostic
CSENSE	100pF	Sense signal filtering
C _{VS}	100nF	Filtering of the voltage spikes on the battery line
C _{OUT}	4.7nF	Protection of the device during ESD and BCI

Table 3. Bill of Materials



Ordering Information (Note 23)

Part Number	Package	Marking	Reel Size (inches)	Tape Width (mm)	Packing	
					Qty.	Carrier
ZXMS81090SP-13	SO-8EP (Type E)	ZXMS81090	13	12	2500	Reel

Note: 23. For packaging details, go to our website at https://www.diodes.com/design/support/packaging/diodes-packaging/.

Marking Information

⊃¦¦ ZXMS81090 YYWW

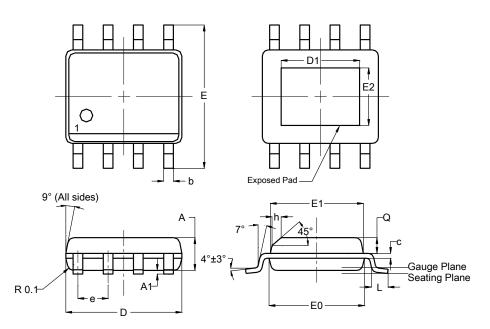
Code Marking ZXMS<u>810</u>90: Product Type Marking Code YY or <u>YY</u>: Year (ex: 23 = 2023) WW or <u>WW</u>: Week 01 to 52; 52 represents week 52 and 53



Package Outline Dimensions

Please see http://www.diodes.com/package-outlines.html for the latest version.

SO-8EP (Type E)

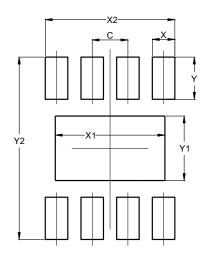


1						
SO-8EP (Type E)						
Dim	Min	Max	Тур			
Α	1.40	1.50	1.45			
A1	0.00	0.13	-			
b	0.30	0.50	0.40			
С	0.15	0.25	0.20			
D	4.85	4.95	4.90			
D1	2.65	3.75	3.70			
Е	5.90	6.10	6.00			
E0	3.85	3.95	3.90			
E1	3.80	3.90	3.85			
E2	2.05	2.15	2.10			
е			1.27			
h	1		0.35			
L	0.62	0.82	0.72			
Ø	0.60	0.70	0.65			
All Dimensions in mm						

Suggested Pad Layout

Please see http://www.diodes.com/package-outlines.html for the latest version.

SO-8EP (Type E)



Dimensions	Value (in mm)	
С	1.270	
Χ	0.802	
X 1	3.900	
X2	4.612	
Υ	1.505	
Y1	2.300	
Y2	6.500	

Mechanical Data

- Moisture Sensitivity: Level 1 per J-STD-020
- Terminals: Finish Matte Tin Plated Leads, Solderable per MIL-STD-202, Method 208 @3
- Weight: 0.08 grams (Approximate)



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